

# BGT24MTR11

Silicon Germanium 24 GHz Transceiver MMIC

## Preliminary Data Sheet

Revision 2.3, 2012-09-19

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**BGT24MTR11 Silicon Germanium 24 GHz Transceiver MMIC**

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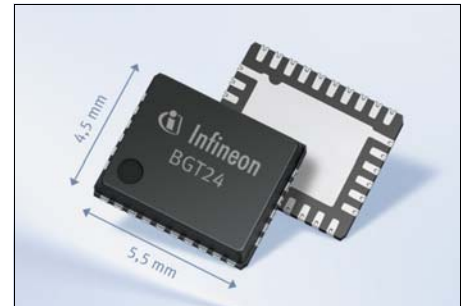
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## 1 Features

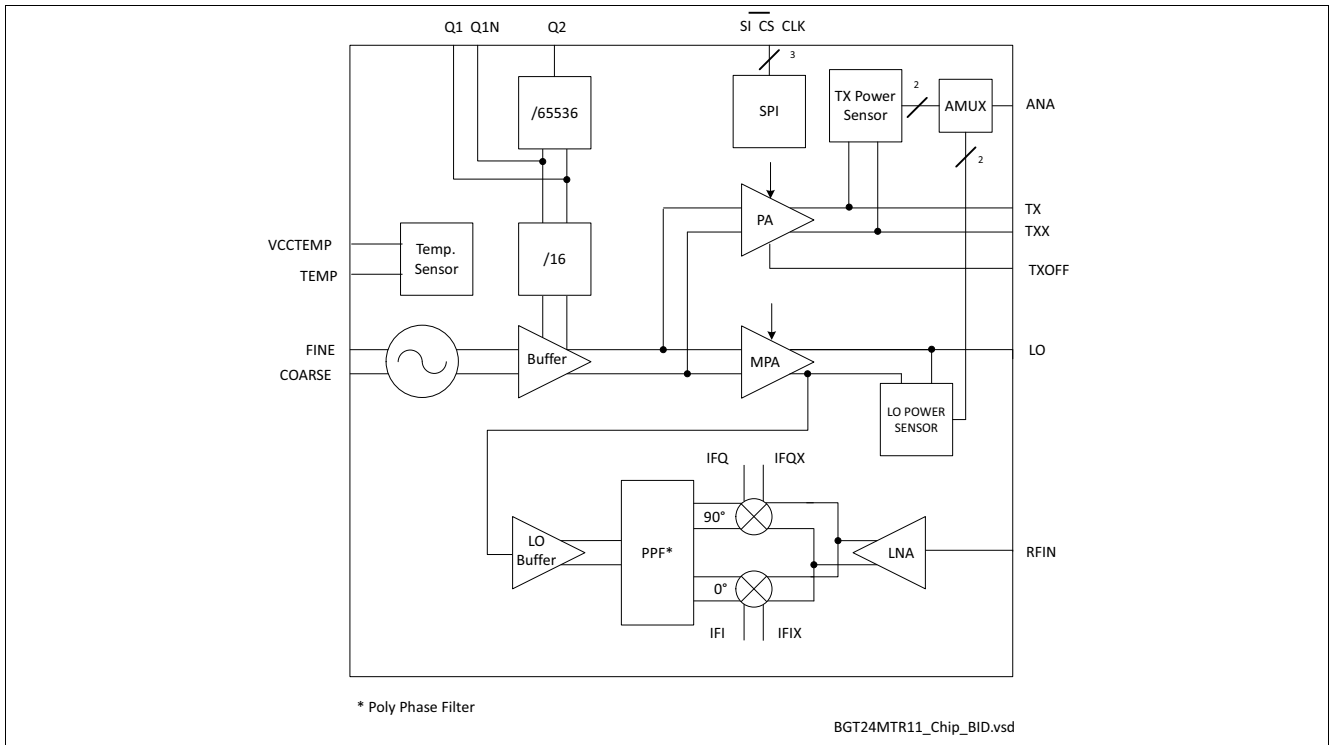
- 24 GHz ISM band transceiver MMIC
- Fully integrated low phase noise VCO
- Switchable prescaler with 1.5 GHz and 23 kHz output
- On chip power and temperature sensors
- Gilbert based homodyne quadrature receiver
- Single ended RF and LO terminal
- Low noise figure  $NF_{SSB}$ : 12 dB
- High conversion gain: 26 dB
- High 1 dB input compression point: -12 dBm
- Single supply voltage 3.3 V
- Low power consumption 500 mW
- 200 GHz bipolar SiGe:C technology b7hf200
- Fully ESD protected device
- VQFN-32-9 leadless plastic package incl. LTI feature
- Pb-free (RoHS compliant) package



### Description

The BGT24MTR11 is a Silicon Germanium MMIC for signal generation and reception, operating from 24.00 to 26.00 GHz. It is based on a 24 GHz fundamental voltage controlled oscillator. Switchable frequency prescalers are included with output frequencies of 1.5 GHz and 23 kHz. The main RF output delivers upto 8dBm signal power to feed an antenna and an auxiliary LO output is available to provide LO signal to separate receiver components. An LO buffer amplifier is included to relax LO drive requirements and a LNA provides low noise figure. A RC polyphase filter (PPF) is used for LO quadrature phase generation of the homodyne quadrature downconversion mixer. Output power sensors as well as a temperature sensor are implemented for monitoring purposes. The device is controlled via SPI and is manufactured in a 0.18 $\mu$ m SiGe:C technology offering a cutoff frequency of 200 GHz. The MMIC is packaged in a 32 pin leadless RoHS compliant VQFN package.

Product Name	Package	Chip	Marking
BGT24MTR11	VQFN32-9	T1524	BGT24MTR11



**Figure 1 BGT24MTR11 Block Diagram**



## 2 Electrical Characteristics

### 2.1 Absolute Maximum Ratings

$T_A = -40\text{ °C}$  to  $105\text{ °C}$ ; all voltages with respect to ground, positive current flowing into pin (unless otherwise specified)<sup>1)</sup>

**Table 1 Absolute Maximum Ratings**

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Supply voltage	$V_{CC} / V_{CCTEMP}$	-0.3	–	3.6	V	–
DC voltage at RF Pins: TX, TXX, LO, RFIN	$VDC_{RF}$	–	–	0	V	MMIC provides short circuit to GND for all RF Pins
DC current into Pin IFI, IFIX, IFQ, IFQX	$I_{IF}$	-6	–	3.5	mA	max. values indicate current due to short circuit to GND and Vcc respectively
DC current into Pin ANA	$I_{ANA}$	-5	–	0.3	mA	max. values indicate current due to short circuit to GND and Vcc respectively
DC current into Pin VCCTEMP	$I_{VCCTEMP}$	-1	–	1.5	mA	max. values indicate current due to short circuit to GND and Vcc respectively
DC current into Pin Q1	$I_{Q1}$	-8	–	8	mA	–
DC current into Pin Q2	$I_{Q2}$	-2.2	–	2.2	mA	–
RF input power into Pin RFIN	$P_{RF}$	–	–	0	dBm	–
DC voltage at Pins Fine, Coarse	$V_{Fine}, V_{Coarse}$	0	–	5	V	–
DC voltage at Pin TXOFF	$VDC_{TXOFF}$	0	–	Vcc+0.3	V	–
Total power dissipation	$P_{DISS}$	–	–	650	mW	–
Junction temperature	$T_J$	-40	–	150	°C	–
Ambient temperature range	$T_A$	-40	–	105	°C	$T_A$ = temperature at package soldering point
Storage temperature range	$T_{STG}$	-40	–	150	°C	–

**Attention: Stresses exceeding the max. values listed here may cause permanent damage to the device. Exposure to absolute maximum rating conditions for extended periods may affect device reliability. Maximum ratings are absolute ratings; exceeding only one of these values may cause irreversible damage to the integrated circuit.**

1) Not subject to production test, specified by design

## 2.2 Thermal Resistance

**Table 2 Thermal Resistance**

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Junction - soldering point <sup>1)</sup>	$R_{thJS}$	–	–	40	K/W	–

1) For calculation of  $R_{thJA}$  please refer to application note thermal resistance

## 2.3 ESD Integrity

**Table 3 ESD Integrity**

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
ESD robustness RFIN, TX, TXX <sup>1)</sup>	$V_{ESD-HBM}$	1000	–	–	V	All RF-pins
ESD robustness low frequency and DC pins <sup>1)</sup>	$V_{ESD-HBM}$	1000	–	–	V	All other pins

1) According to ESDA/JEDEC Joint Standard for Electrostatic Discharge Sensitivity Testing, Human Body Model (HBM)-Component Level, ANSI/ESDA/JEDEC JS-001-2011  
Please note that this result is subject to  
- lot variations within the manufacturing process as specified by Infineon  
- changes in the specific test setup

## 2.4 Measured RF Characteristics

### 2.4.1 Power Supply

**Table 4** Typical Characteristics  $T_A = -40 \dots 105 \text{ }^\circ\text{C}$

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Supply voltage	$V_{CC}$	3.135	3.3	3.465	V	–
Supply current	$I_{CC}$	120	150	180	mA	–
Supply voltage temperature sensor	$V_{CCTemp}$	3.135	3.3	3.465	V	–
Supply current temperature sensor	$I_{CCTemp}$	–	2.2	–	mA	–

### 2.4.2 TX Section

**Table 5** Typical Characteristics  $T_A = -40 \dots 105 \text{ }^\circ\text{C}^1)$

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
VCO frequency range	$f_{VCO}$	24.00	–	26.00	GHz	–
VCO tuning voltage for VCO frequency range	$V_{FINE}, V_{COARSE}$	0.5 <sup>2)</sup>	–	5.0 <sup>3)</sup>	V	–
VCO tuning slope	$\Delta f / \Delta V_{FINE}$	–	–	1000	MHz/V	–
VCO tuning slope	$\Delta f / \Delta V_{COARSE}$	–	–	1800	MHz/V	–
VCO pushing	$\Delta f / \Delta V_{CC}$	–	–	300	MHz/V	@ f = 24 GHz
VCO phase noise	$P_N$	–	-85	-75	dBc/Hz	@ 100 kHz offset
TX/TXX load impedance <sup>4)</sup>	$Z_{TXLOAD}$	–	100	–	$\Omega$	With off chip compensation network as proposed
Max. TX output power	$P_{TX}$	1	8	12	dBm	–
TX output power adjustable range	$a_{TX}$	3	9	–	dB	Adjustable via SPI
TX output power in “off” mode	$P_{TXoff}$	–	–	-15	dBm	–
LO load impedance <sup>4)</sup>	$Z_{LOLOAD}$	–	50	–	$\Omega$	With off chip compensation network as proposed
LO output power	$P_{LO}$	-8	0	6	dBm	–
Q1 Prescaler division ratio	$D_{Q1}$	–	16	–	–	–
Q1 Prescaler output power	$P_{Q1}$	-12	-9	-6	dBm	Q1 loaded with 100 Ohm (AC-coupled)
Q2 Prescaler division ratio	$D_{Q2}$	–	1048576	–	–	–

**Table 5** Typical Characteristics  $T_A = -40 \dots 105 \text{ }^\circ\text{C}^{1)}$  (cont'd)

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Q2 Prescaler max. output voltage	$V_{\max Q2}$	$V_{CC}-0.7$	–	–	V	Test condition: Q2 loaded with high impedance probe (1 MOhm, 13 pF)
Q2 Prescaler min. output voltage	$V_{\min Q2}$	–	–	0.5	V	Test condition: Q2 loaded with high impedance probe (1 MOhm, 13 pF)
Q2 Prescaler max. output source current	$I_{\max source Q2}$	2	–	–	mA	Test condition: Q2 loaded with 50 Ohm to $V_{CC}$
Q2 Prescaler max. output sink current	$I_{\max sink Q2}$	2	–	–	mA	Test condition: Q2 loaded with 50 Ohm to $V_{CC}$
Q2 Prescaler output resistance in disable mode	$R_{Q2,DIS}$	100	–	–	k $\Omega$	–
Voltage at Txoff for disabling TX output power	$V_{TX,OFF}$	1.5	–	–	V	–
Voltage at Txoff for enabling TX output power	$V_{TX,ON}$	–	–	0.5	V	–
TXon/off switching time	$t_{ON/OFF}$	–	–	500	ns	–

1) Performance based on application circuit in Figure 2 on Page 15

2) Min. limit @ 25°C = 0.8V; min. limit @ 105°C = 1.15V

3) Max. limit for max. frequency of 24.25GHz = 3.1V; max. limit for max. frequency of 24.5GHz = 3.8V

4) Guaranteed by device design

### 2.4.3 RX Section

**Table 6** Typical Characteristics  $T_A = -40 \dots 105 \text{ }^\circ\text{C}^{1)}$

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
RFIN frequency range	$f_{RFIN}$	24.00	–	26.00	GHz	–
RFIN port impedance <sup>2)</sup>	$Z_{RFIN}$	–	50	–	$\Omega$	With off chip compensation network as proposed
RFIN VSWR	$VSWR$	–	–	2:1	–	With off chip compensation network as proposed
IF frequency range	$f_{IF}$	0	–	1	MHz	–
IF output impedance <sup>2)</sup>	$Z_{IF}$	–	800	–	$\Omega$	–
Voltage conversion gain	$G_C$	18	26	33	dB	$R_{LOAD,IF} > 10 \text{ k}\Omega$

**Table 6** Typical Characteristics  $T_A = -40 .. 105\text{ °C}^1$  (cont'd)

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
SSB noise figure	$N_{SSB}$	–	12	20	dB	Single sideband at $f_{IF} = 100\text{ kHz}$
IF 1/f corner frequency	$f_c$	–	10	20	kHz	–
Input compression point	$IP_{1dB}$	-18	-12	–	dBm	–
Input 3'rd order intercept point	$IIP3$	-8	-4	–	dBm	–
Quadrat. phase imbalance	$\varepsilon_p$	–	–	10	deg	–
Quadrat. amplitude imbalance	$\varepsilon_A$	–	–	1	dB	–

- 1) Performance based on application circuit in Figure 2 on Page 15
- 2) Guaranteed by device design

## 2.5 Temperature Sensor

Monitoring of the chip temperature is provided by the on-chip temperature sensor which delivers temperature-proportional voltage to the TEMP output pad. The temp. sensor can be independently biased through VCCTEMP. Thereby the chip temperature can be monitored while the main supply of the Transceiver is switched off.

**Table 7 Typical Characteristics Temperature Sensor  $T_A = -40 .. 105\text{ }^\circ\text{C}$ ,  $V_{CC} = 3.3\text{ V}^1$**

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Temperature range	$T_{\text{TSENS}}$	-40	–	105	$^\circ\text{C}$	–
Output temperature voltage	$V_{\text{TOFFS}}$	–	1.55	–	V	@ 25 $^\circ\text{C}$
Sensitivity	$S_{\text{TSENS}}$	3	4	–	mV/K	–

1) all voltages with respect to ground, positive current flowing into pin (unless otherwise specified)

## 2.6 Power Detector

For output power indication, peak voltage detectors are connected to the output of the TX power amplifier and to the LO medium power amplifier. To eliminate temperature and supply voltage variations, a reference output voltage  $V_{\text{REF}}$  is available through the ANA output for the TX and LO power sensor. The compensated detector output voltage is given by the difference between  $V_{\text{OUT}}$  and  $V_{\text{REF}}$  for both power sensors respectively. This voltage is proportional to the RF voltage swing at the individual amplifier outputs, its characteristic is non-directional.

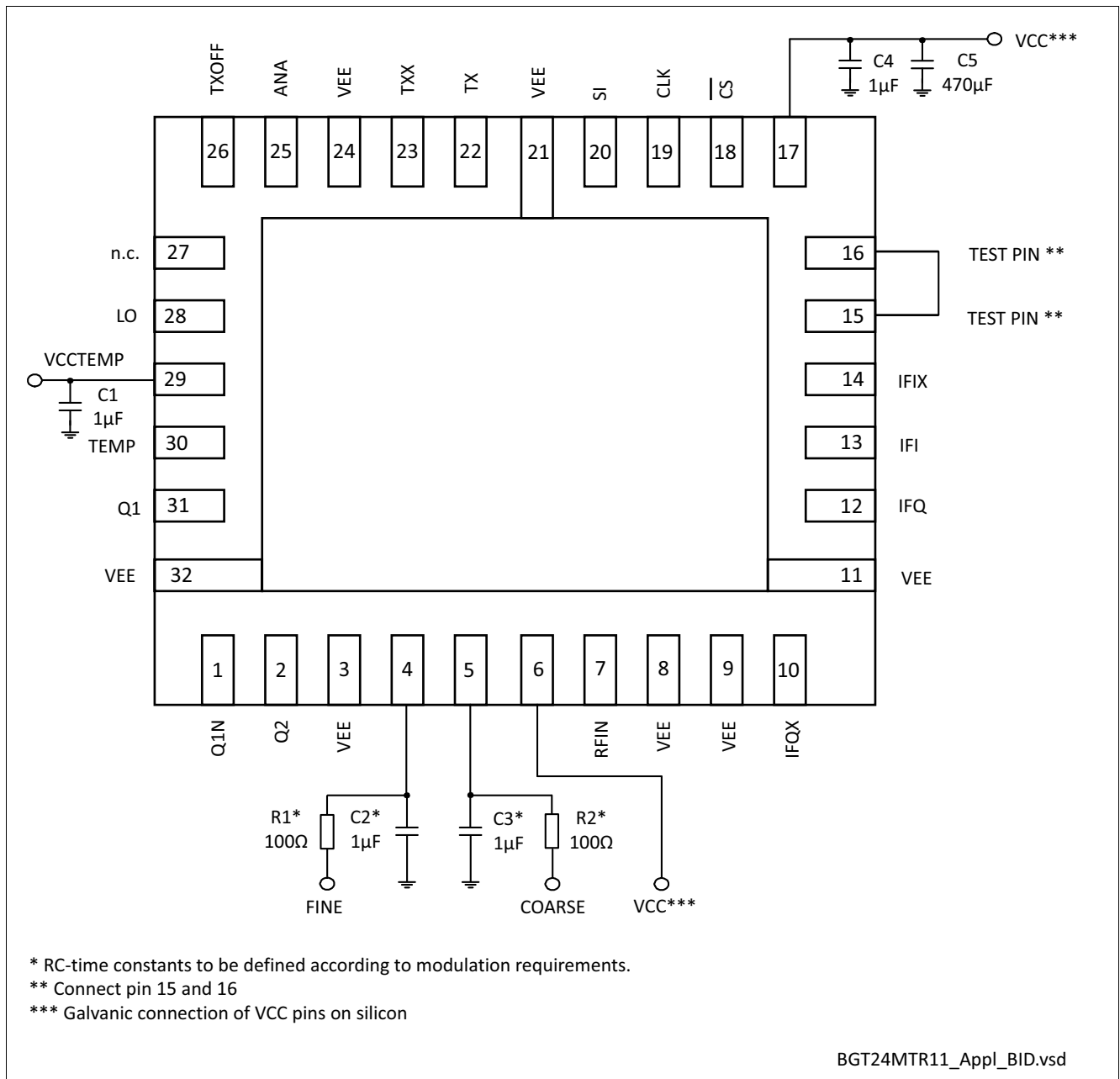
**Table 8 Typical Characteristics Power Detector  $T_A = -40 .. 105\text{ }^\circ\text{C}$ ,  $V_{CC} = 3.3\text{ V}^1$**

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Power range	$P_{\text{PSENS}}$	-10	–	12	dBm	–
TX power sensor output	$V_{\text{OUT,TX}} - V_{\text{REF,TX}}$	–	500	–	mV	@ $P_{\text{TX}} = 8\text{ dBm}$
LO power sensor output	$V_{\text{OUT,LO}} - V_{\text{REF,LO}}$	–	50	–	mV	@ $P_{\text{LO}} = 0\text{ dBm}$

1) all voltages with respect to ground, positive current flowing into pin (unless otherwise specified)

### 3 Application Circuit and Block Diagram

#### 3.1 Application Circuit Schematic



**Figure 2 Application Circuit with Chip Outline (Top View)**

**Table 9 Bill of Materials**

Part Number	Part Type	Manufacturer	Size	Comment
C1 ... C5	Chip capacitor	Various	Various	–
R1 ... R2	Chip resistor	Various	0402	–

### 3.2 Pin Description

**Table 10 Pin Definition and Function**

Pin No.	Name	Function
1	Q1N	Complementary prescaler output 1.5GHz
2	Q2	Prescaler output 23kHz
3	VEE	Ground
4	FINE	VCO fine tuning input
5	COARSE	VCO coarse tuning input
6	VCC	Supply voltage; Total current divided equal on both VCC pins
7	RFIN	RF input downconverter
8	VEE	Ground
9	VEE	Ground
10	IFQX	Complementary quadrature phase IF output downconverter
11	VEE	Ground
12	IFQ	Quadrature phase IF output downconverter
13	IFI	In phase IF output downconverter
14	IFIX	Complementary in phase IF output downconverter
15	TEST PIN	Test pin; DC coupled pin
16	TEST PIN	Test pin; DC coupled pin
17	VCC	Supply voltage; Total current divided equal on both VCC pins
18	$\overline{CS}$	Chip select input SPI (inverted)
19	CLK	Clock input SPI interface
20	SI	Data input SPI interface
21	VEE	Ground
22	TX	Transmit output
23	TXX	Complementary transmit output
24	VEE	Ground
25	ANA	Analog output
26	TXOFF	Pulsable Pin / Please connect to VEE in case TXOFF function is controlled via SPI
27	n.c.	Not connected
28	LO	LO output
29	VCCTEMP	Temperature sensor supply voltage
30	TEMP	Temperature sensor output
31	Q1	Prescaler output 1.5GHz
32	VEE	Ground



### 3.3 SPI

Communication to the transceiver is done via a Serial-Peripheral-Interface (SPI). The 16 bit SPI has a hardwired Power-On reset, which sets the output bits to a defined state after turning on the supply voltage. Data transmission is started by a negative edge on  $\overline{CS}$ . Data at SI is then read at the falling edge of CLK. The most significant bit (MSB) is read first.

**Table 11 SPI Data Bit Description**

Data Bit	Name	Description (Logic High)	Power ON State
15 (MSB)	GS	LNA Gain reduction	low
14 ..13	–	Not used	low
12	DIS_PA	TX power disabled, in case TXon/off function is controlled via TXOFF pin, this bit needs to be set in low state	high
11	AMUX2	Analog multiplexer control bit 2	high
10	Test Bit	Test bit, must be low otherwise malfunction	low
9	Test Bit	Test bit, must be low otherwise malfunction	low
8	AMUX1	Analog multiplexer control bit 1	low
7	AMUX0	Analog multiplexer control bit 0	low
6	DIS_DIV64k	Disable 64k divider	low
5	DIS_DIV16	Disable 16 divider	low
4	PC2_BUF	High LO buffer output power	low
3	PC1_BUF	High TX buffer output power	low
2	PC2_PA	TX power reduction bit 2	high
1	PC1_PA	TX power reduction bit 1	high
0	PC0_PA	TX power reduction bit 0	high

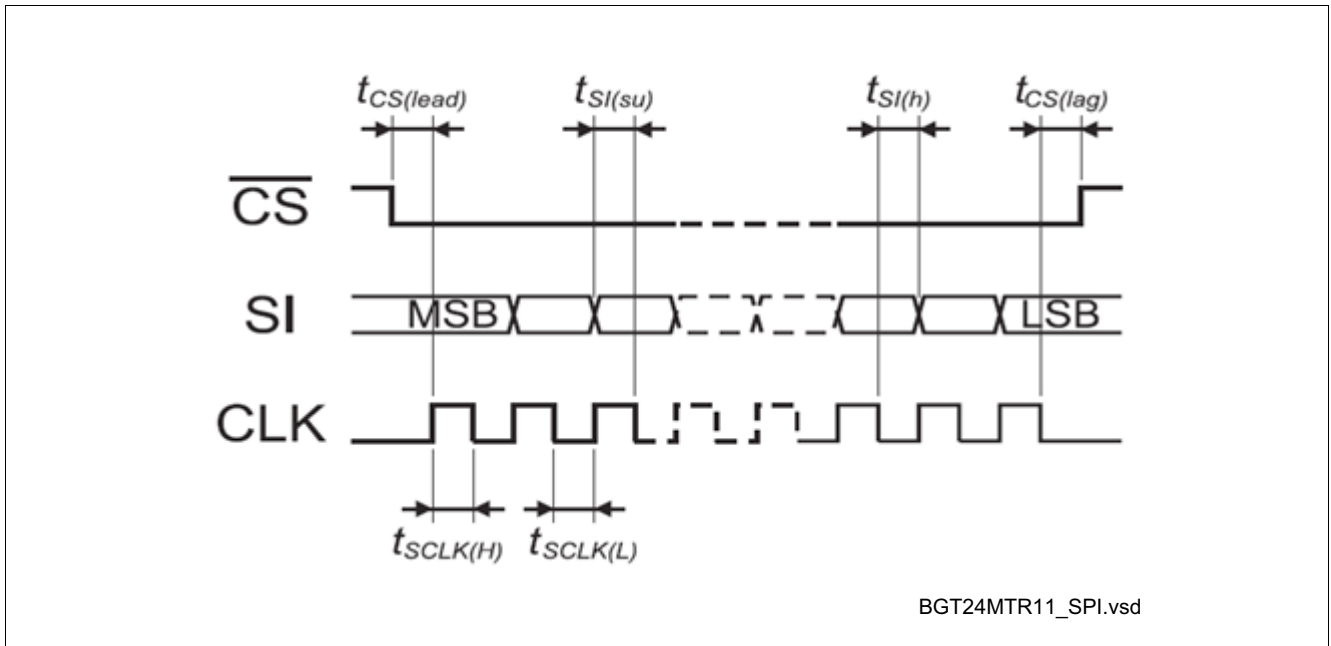


Figure 3 Timing Diagram of the SPI

Table 12 SPI Timing and Logic Levels

Parameter	Symbol	Values			Unit
		Min.	Typ.	Max.	
Serial clock frequency	$f_{SCLK}$	0	–	50	MHz
Serial clock high time	$f_{SCLK(H)}$	10	–	–	ns
Serial clock low time	$t_{SCLK(L)}$	10	–	–	ns
Chip select lead time	$t_{CS(lead)}$	20	–	–	ns
Chip select lag time	$t_{CS(lag)}$	20	–	–	ns
Data setup time	$t_{SI(su)}$	10	–	–	ns
Data hold time	$t_{SI(h)}$	10	–	–	ns
Low level (SI, CLK, $\overline{CS}$ )	$V_{IN(L)}$	0	–	0.8	V
High level (SI, CLK, $\overline{CS}$ )	$V_{IN(H)}$	2.0	–	$V_{CC} + 0.3V$	V
Input current	$I_{IN}$	-150	–	150	$\mu A$

Table 13 Truth Table AMUX

Output signal ANA	AMUX2	AMUX1	AMUX0
$P_{OUT,TX}$	low	low	low
$P_{REF,TX}$	low	low	high
$P_{OUT,LO}$	low	high	low
$P_{REF,LO}$	low	high	high
$V_{TEMP}$	high	low	low
Test_Signal1	high	low	high

**Table 13 Truth Table AMUX (cont'd)**

<b>Output signal ANA</b>	<b>AMUX2</b>	<b>AMUX1</b>	<b>AMUX0</b>
Test_Signal2	high	high	low
Test_Signal2	high	high	high

### 3.4 Application Board and Reflow Profile

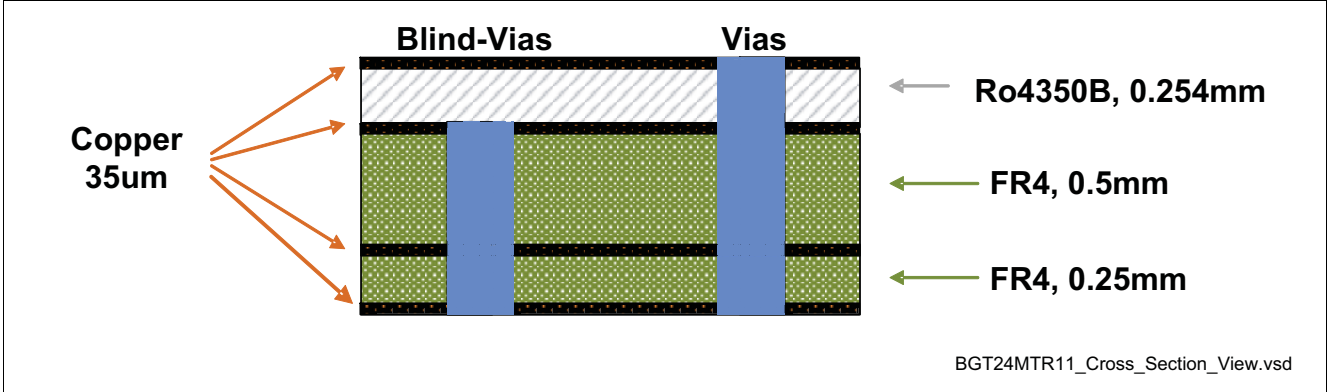


Figure 4 Cross-Section View of Application Board

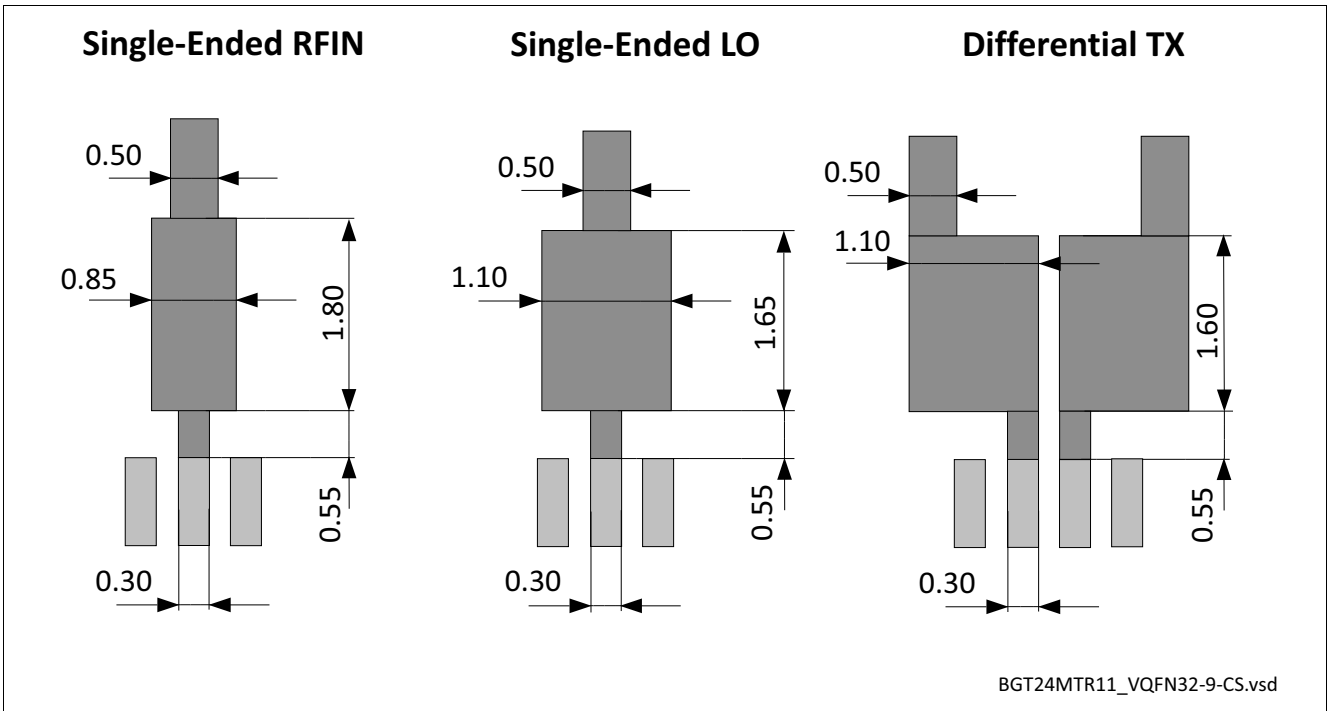


Figure 5 Detail of Compensation Structure

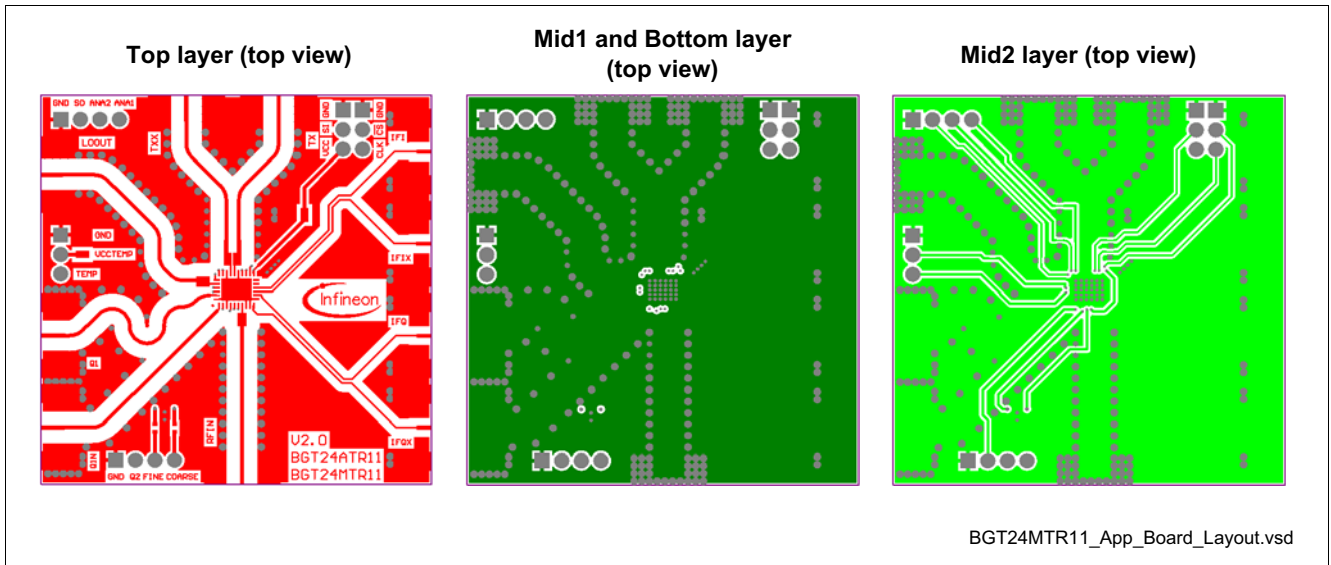


Figure 6 Application Board Layout

Note: In order to achieve the same performance as given in this datasheet please follow the suggested PCB-layout as closely as possible. The compensation structure is critical for RF performance. Via holes as recommended on next page (not shown above).

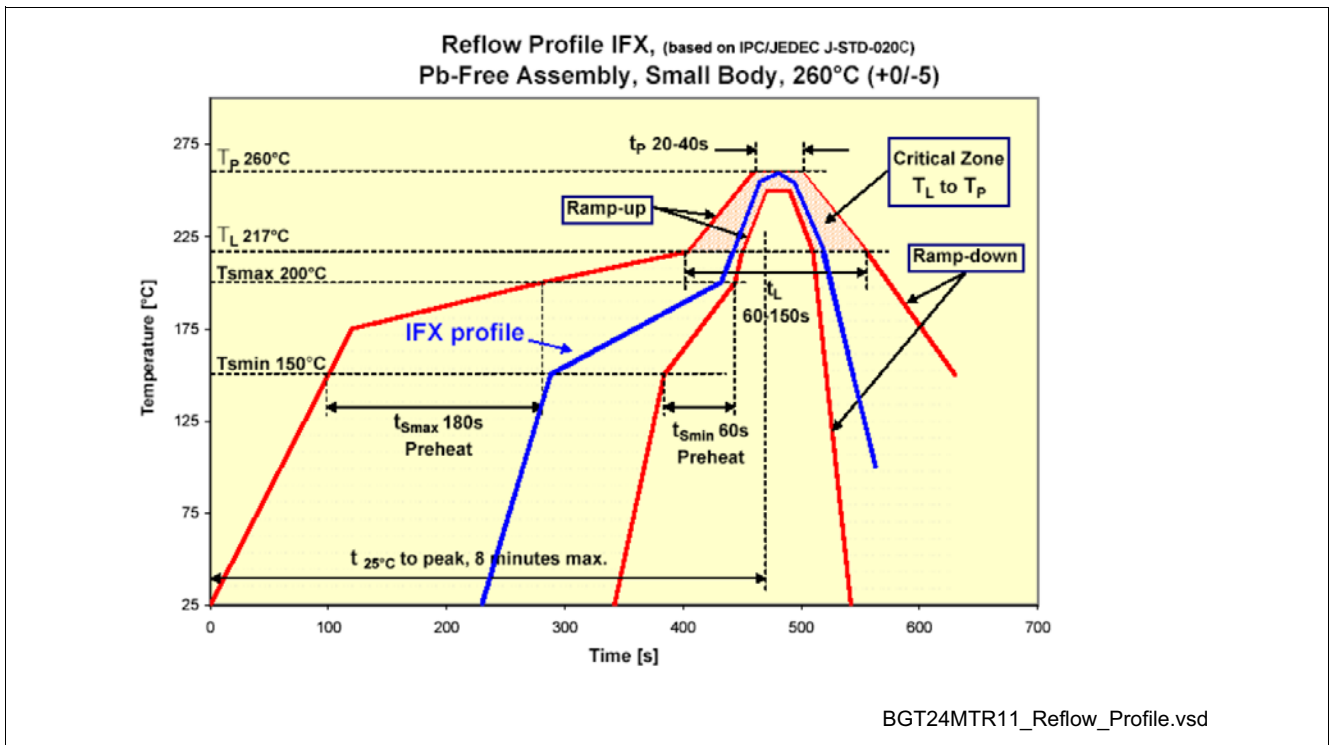
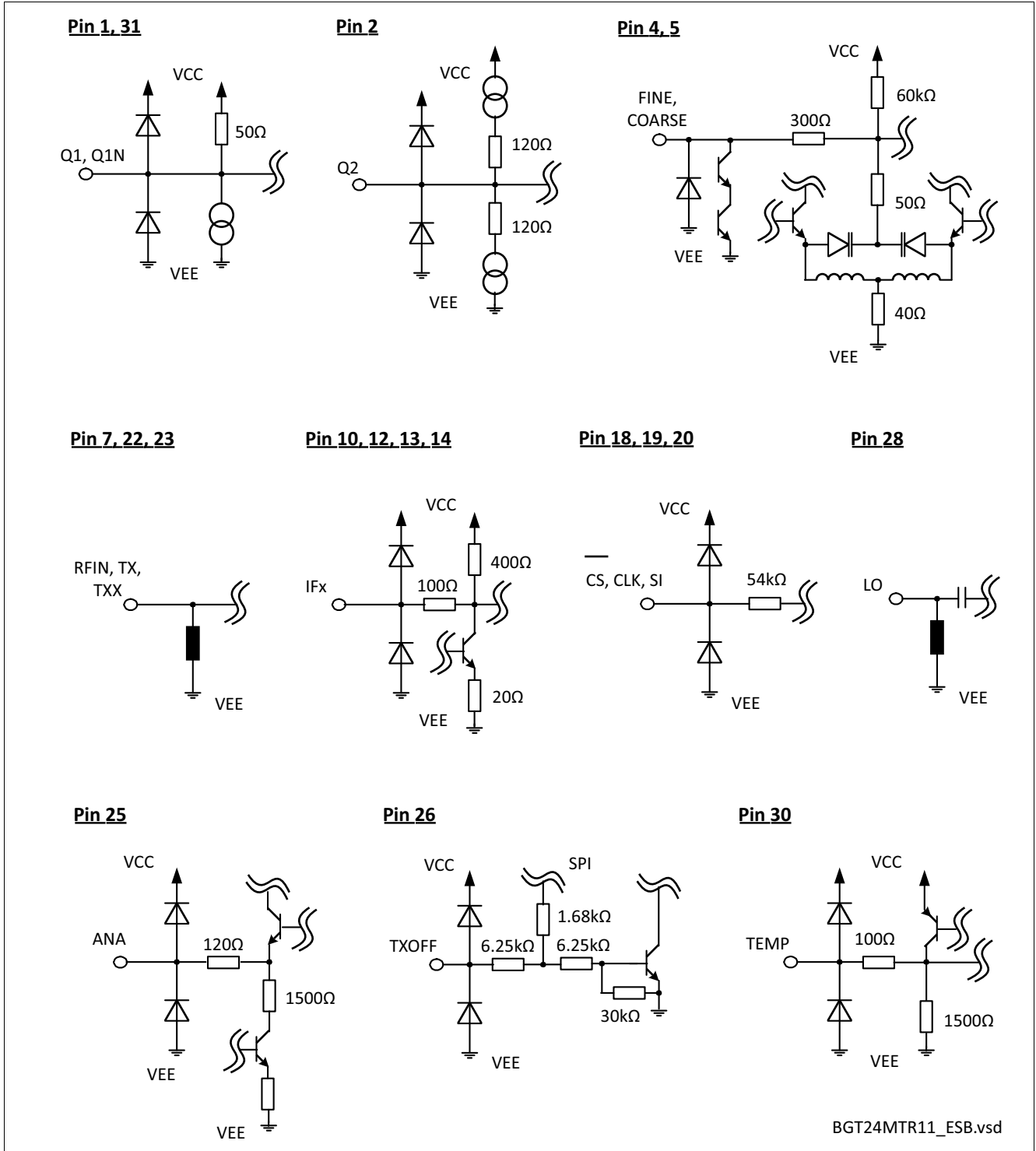


Figure 7 Reflow Profile for BGT24MTR11 (VQFN32-9)

### 3.5 Equivalent Circuit Diagram of MMIC Interfaces



**Figure 8** Equivalent Circuit Diagram of MMIC Interfaces

## 4 Physical Characteristics

### 4.1 Package Footprint

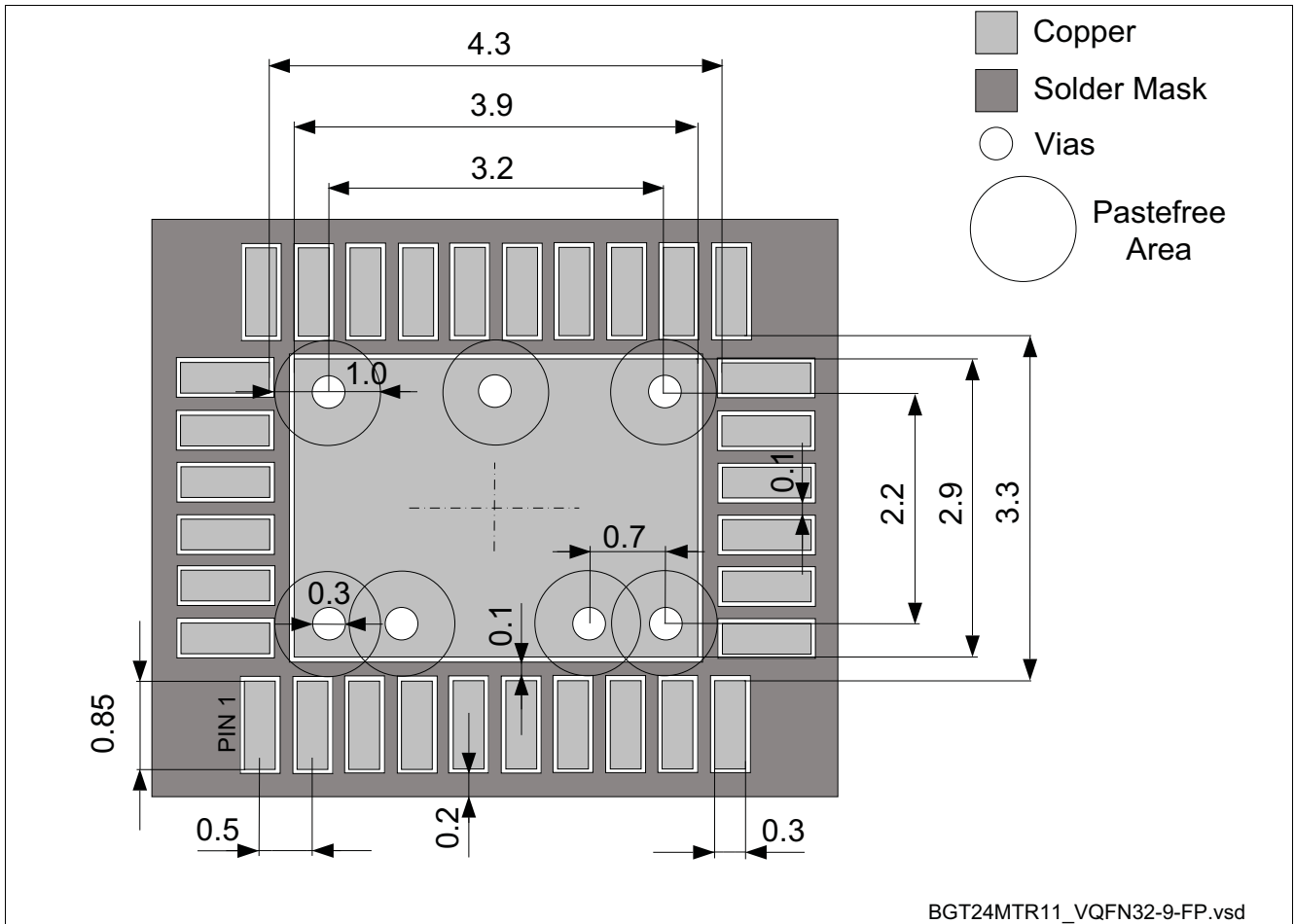


Figure 9 Recommended Footprint and Stencil Layout for the VQFN32-9 Package

4.2 Package Dimensions

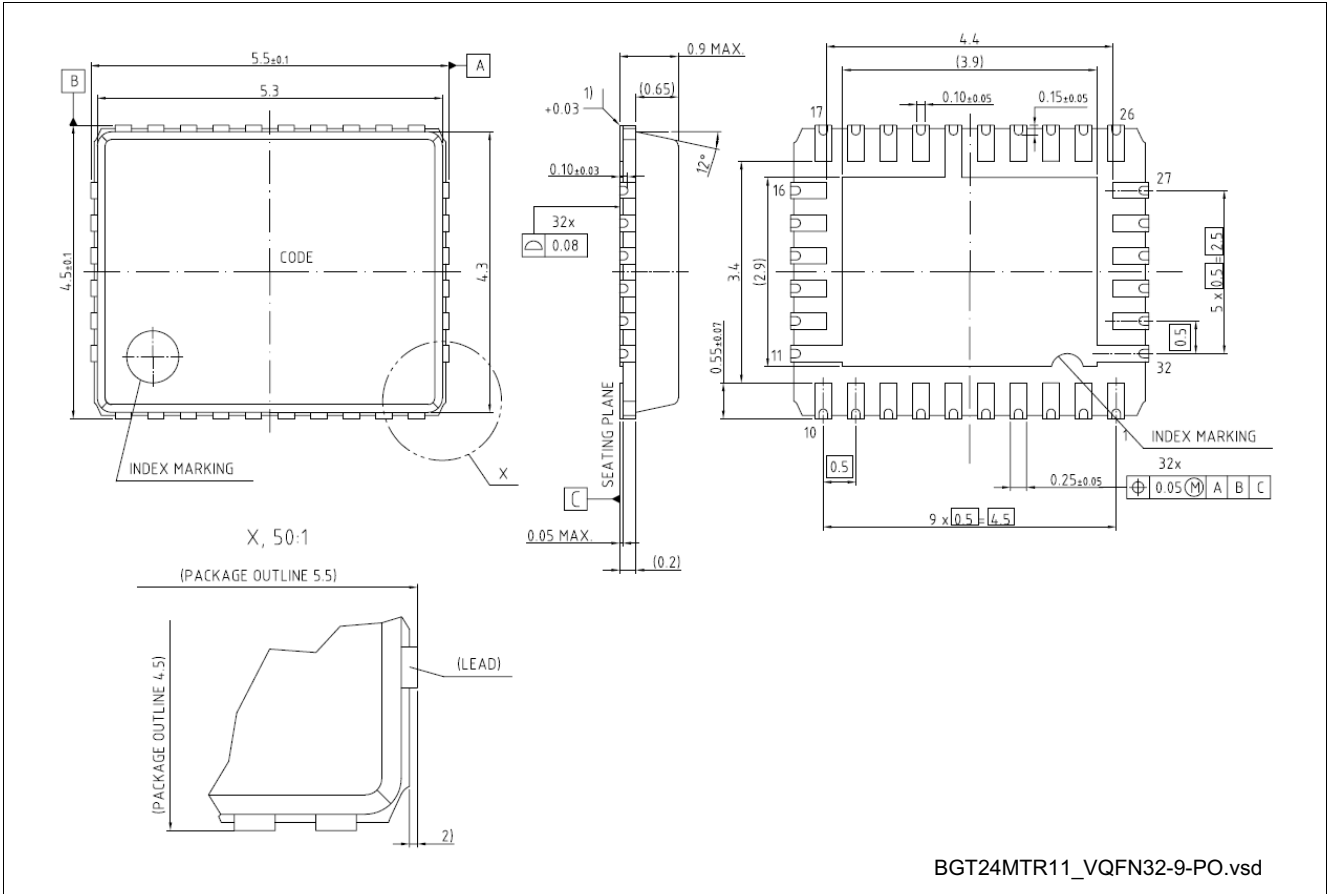


Figure 10 Package Outline (Top, Side and Bottom View) of VQFN32-9

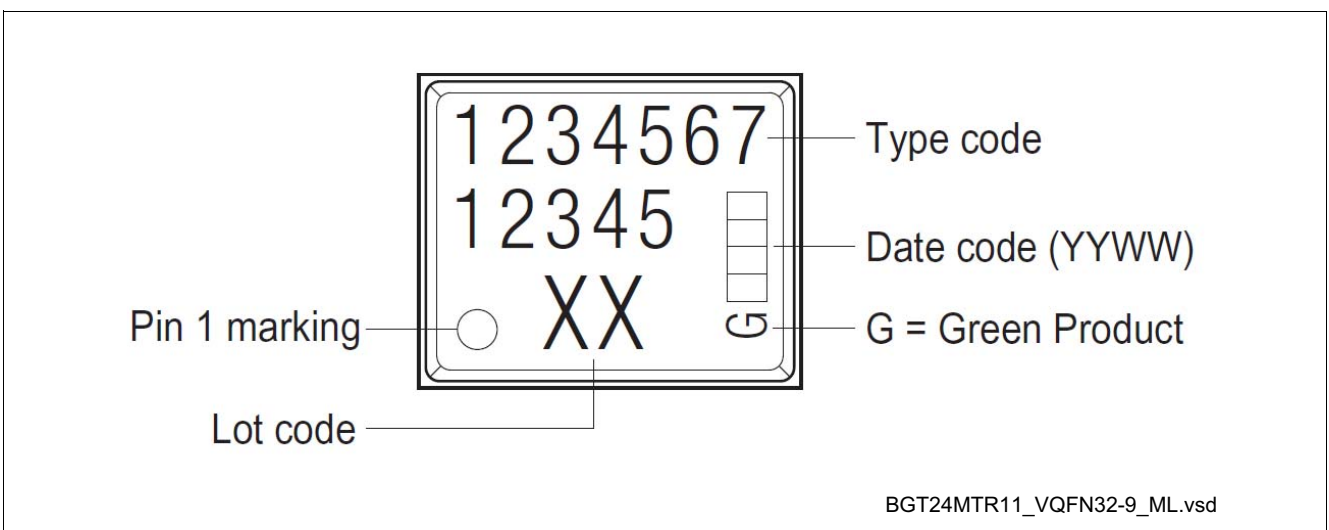
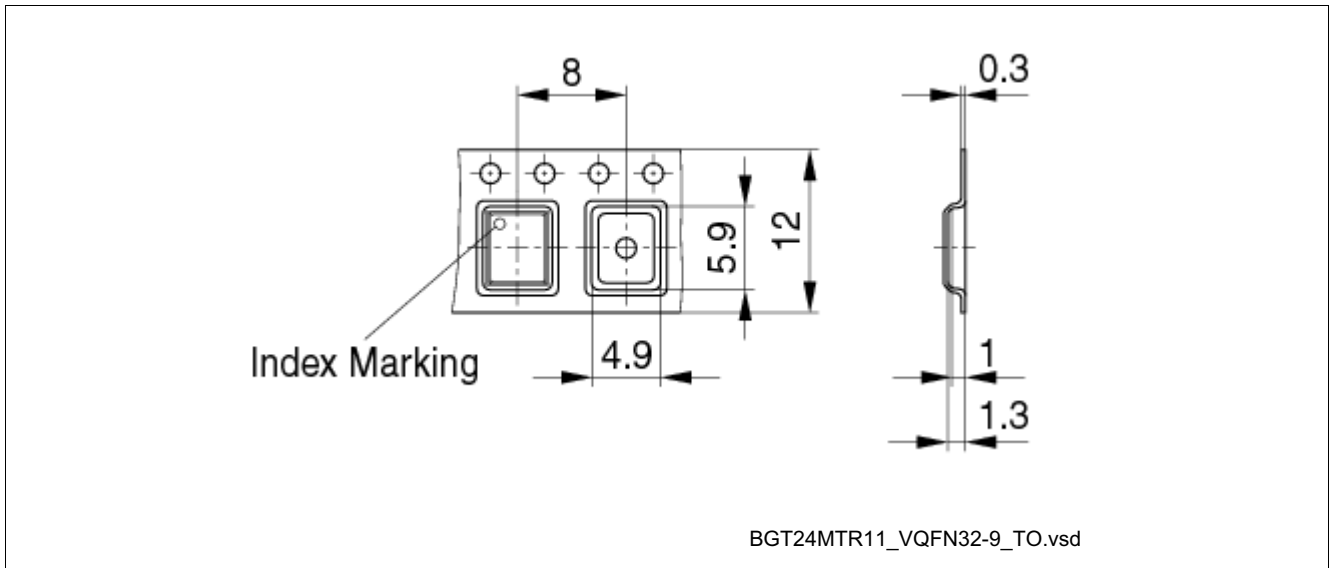


Figure 11 Marking Layout VQFN32-9





**Figure 12** Tape of VQFN32-9

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