

2.5A POWER SWITCHING REGULATOR

- 2.5A OUTPUT CURRENT
- 5.1V TO 40V OPUTPUT VOLTAGE RANGE
- PRECISE (± 2%) ON-CHIP REFERENCE
- HIGH SWITCHING FREQUENCY
- VERY HIGH EFFICIENCY (UP TO 90%)
- VERY FEW EXTERNAL COMPONENTS
- SOFT START
- INTERNAL LIMITING CURRENT
- THERMAL SHUTDOWN

HEPTAWATT ORDERING NUMBERS: L4960 (Vertical) L4960H (Horizontal)

DESCRIPTION

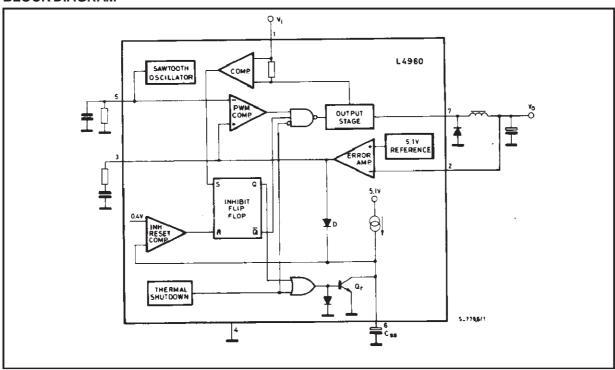
The L4960 is a monolithic power switching regulator delivering 2.5A at a voltage variable from 5V to 40V in step down configuration.

Features of the device include current limiting, soft start, thermal protection and 0 to 100% duty cycle for continuous operation mode.

The L4960 is mounted in a Heptawattplastic power package and requires very few external components.

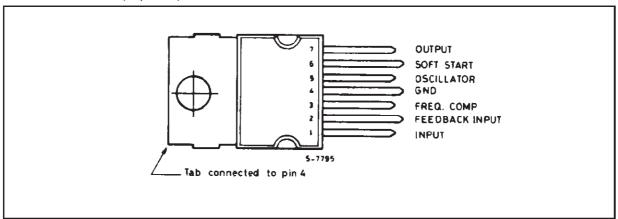
Efficient operation at switching frequencies up to 150KHz allows a reduction in the size and cost of external filter components.

BLOCK DIAGRAM



June 2000 1/16

PIN CONNECTION (Top view)



ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
V ₁	Input voltage	50	V
V ₁ - V ₇	Input to output voltage difference	50	V
V ₇	Negative output DC voltage	-1	V
	Negative output peak voltage at t = 0.1μs; f = 100KHz	-5	V
V ₃ , V ₆	Voltage at pin 3 and 6	5.5	V
V ₂	Voltage at pin 2	7	V
l ₃	Pin 3 sink current	1	mA
l ₅	Pin 5 source current	20	mA
P _{tot}	Power dissipation at T _{case} ≤ 90°C	15	W
T _j , T _{stg}	Junction and storage temperature	-40 to 150	°C

PIN FUNCTIONS

N°	NAME	FUNCTION
1	SUPPLY VOLTAGE	Unregulated voltage input. An internal regulator powers the internal logic.
2	FEEDBACK INPUT	The feedback terminal of the regulation loop. The output is connected directly to this terminal for 5.1V operation; it is connected via a divider for higher voltages.
3	FREQUENCY COMPENSATION	A series RC network connected between this terminal and ground determines the regulation loop gain characteristics.
4	GROUND	Common ground terminal.
5	OSCILLATOR	A parallel RC network connected to this terminal determines the switching frequency.
6	SOFT START	Soft start time constant. A capacitor is connected between this terminal and ground to define the soft start time constant. This capacitor also determines the average short circuit output current.
7	ОИТРИТ	Regulator output.

THERMAL DATA

Symbol	Parameter	Value	Unit
R _{th j-case}	Thermal resistance junction-case max	4	°C/W
R _{th j-amb}	Thermal resistance junction-ambient max	50	°C/W

ELECTRICAL CHARACTERISTICS (Refer to the test circuit, $T_j = 25$ °C, $V_i = 35V$, unless otherwise specified)

Symbol	Parameter	Test Co	onditions	Min.	Тур.	Max.	Unit
DYNAMI	C CHARACTERISTICS						
Vo	Output voltage range	V _i = 46V	I _o = 1A	V _{ref}		40	V
Vi	Input voltage range	$V_0 = V_{ref}$ to 36V	I _o = 2.5A	9		46	V
ΔVo	Line regulation	$V_i = 10V \text{ to } 40V$ V_i	$_{o} = V_{ref}$ $I_{o} = 1A$		15	50	mV
ΔVo	Load regulation	$V_o = V_{ref}$	$I_0 = 0.5A \text{ to } 2A$		10	30	mV
V _{ref}	Internal reference voltage (pin 2)	V _i = 9V to 46V	I ₀ = 1A	5	5.1	5.2	V
$\frac{\Delta V_{ref}}{\Delta T}$	Average temperature coefficient of refer voltage	$T_j = 0$ °C to 125°C $I_o = 1$ A			0.4		mV/°C
V _d	Dropout voltage	I ₀ = 2A			1.4	3	V
I _{om}	Maximum operating load current	V _i = 9V to 46V V _o = V _{ref} to 36V		2.5			А
I _{7L}	Current limiting threshold (pin 7)	$V_i = 9V \text{ to } 46V$ $V_o = V_{\text{ref}} \text{ to } 36V$		3		4.5	А
I _{SH}	Input average current	V _i = 46V; output	short-circuit		30	60	mA
η	Efficiency	f = 100KHz	$V_o = V_{ref}$		75		%
		I ₀ = 2A	V _o = 12V		85		%
SVR	Supply voltage ripple rejection	$\begin{array}{l} \Delta \ V_i = 2 V_{rms} \\ \text{fripple} = 100 Hz \\ V_o = V_{ref} \end{array}$	lo = 1A	50	56		dB
f	Switching frequency			85	100	115	KHz
$\frac{\Delta f}{\Delta V_i}$	Voltage stability of switching frequency	V _i = 9V to 46V			0.5		%
$\frac{\Delta f}{\Delta T_j}$	Temperature stability of switching frequency	$T_j = 0$ °C to 125°C			1		%
f _{max}	Maximum operating switching frequency	$V_{o} = V_{ref}$	I ₀ = 2A	120	150		KHz
T _{sd}	Thermal shutdown junction temperature				150		°C

ELECTRICAL CHARACTERISTICS (continued)

I _{1Q}	RACTERISTICS Quiescent drain current	100% duty cycle						
I _{1Q}	Quiescent drain current							
		pins 5 and 7 open	\/ 46\/		30	40	mA	
		0% duty cycle	$V_i = 46V$		15	20	mA	
-l _{7L}	Output leakage current	0% duty cycle				1	mA	
SOFT ST	ART							
I _{6SO}	Source current			100	140	180	μА	
I _{6SI}	Sink current			50	70	120	μА	
RROR A	AMPLIFIER							
V _{3H}	High level output voltage	V ₂ = 4.7V	I ₃ = 100μA	3.5			V	
V _{3L}	Low level output voltage	V ₂ = 5.3V	I ₃ = 100μA			0.5	V	
I _{3SI}	Sink output current	V ₂ = 5.3V		100	150		μΑ	
-l _{3SO}	Source output current	V ₂ = 4.7V		100	150		μА	
l ₂	Input bias current	V ₂ = 5.2V			2	10	μΑ	
Gv	DC open loop gain	V ₃ = 1V to 3V		46	55		dB	
OSCILLATOR								
-l ₅	Oscillator source current			5			mA	

CIRCUIT OPERATION (refer to the block diagram)

The L4960 is a monolithic stepdownswitching regulator providing output voltages from 5.1V to 40V and delivering 2.5A.

The regulation loop consists of a sawtooth oscillator, error amplifier, comparator and the output stage. An error signal is produced by comparing the output voltage with a precise 5.1V on-chip reference (zener zap trimmed to \pm 2%).

This error signal is then compared with the sawtooth signal to generate the fixed frequency pulse width modulated pulses which drive the output stage.

The gain and frequency stability of the loop can be adjusted by an external RC network connected to pin 3. Closing the loop directly gives an output voltage of 5.1V. Higher voltages are obtained by inserting a voltage divider.

Output overcurrents at switch on are prevented by the soft start function. The error amplifier output is initially clamped by the external capacitor C_{ss} and

allowed to rise, linearly, as this capacitor is charged by a constant current source. Output overload protection is provided in the form of a current limiter. The load current is sensed by an internal metal resistor connected to a comparator. When the load current exceeds a preset threshold this comparator sets a flip flop which disables the output stage and discharges the soft start capacitor. A second comparator resets the flip flop when the voltage across the soft start capacitor has fallen to 0.4V.

The output stage is thus re-enabled and the output voltage rises under control of the soft start network. If the overload condition is still present the limiter will trigger again when the threshold current is reached. The average short circuit current is limited to a safe value by the dead time introduced by the soft start network. The thermal overload circuit disables circuit operation when the junction temperature reaches about 150°C and has hysteresis to prevent unstable conditions.

Figure 1. Soft start waveforms

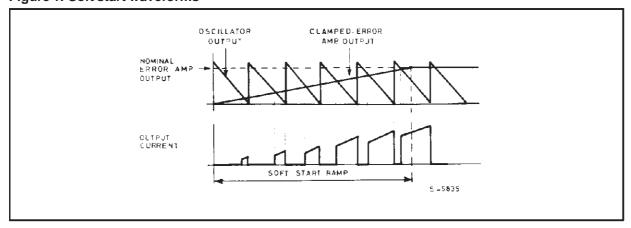
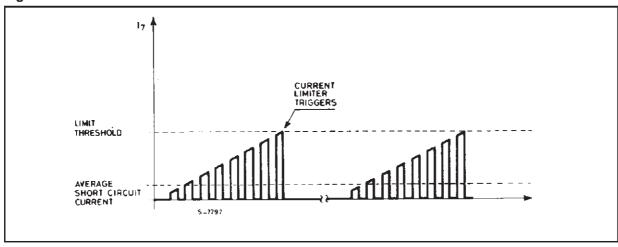


Figure 2. Current limiter waveforms



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Figure 3. Test and application circuit

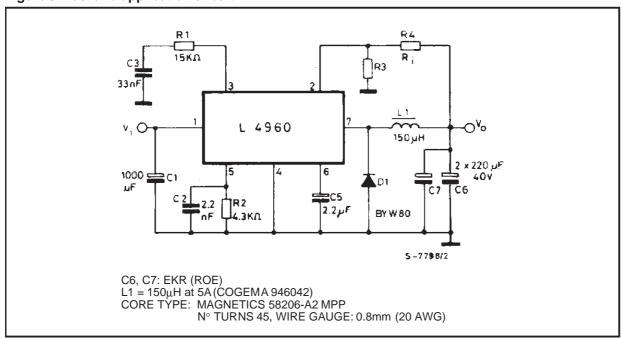


Figure 4. Quiescent drain current vs. supply voltage (0% duty cycle)

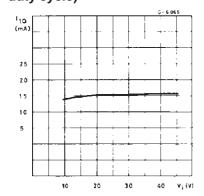


Figure 5. Quiescent drain current vs. supply voltage (100% duty cycle)

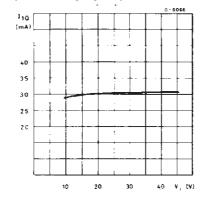


Figure 6. Quiescent drain current vs. junction temperature (0% duty cycle)

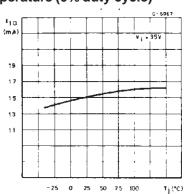


Figure 7. Quiescent drain current vs. junction temperature (100% duty cycle)

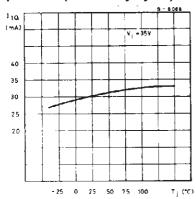


Figure 8. Reference voltage (pin 2) vs. Vi

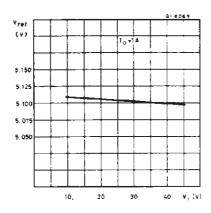


Figure 9. Reference voltage versus junction temperature (pin 2)

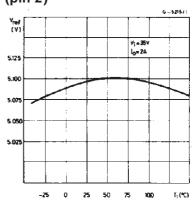


Figure 10. Open loop frequency and phase responde of error amplifier

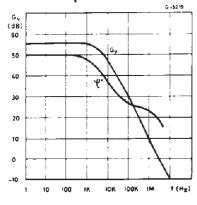


Figure 11. Switching frequency vs. input voltage

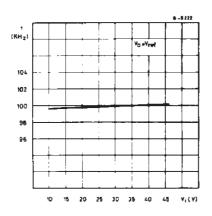


Figure 12. Switching frequency vs. junction temperature

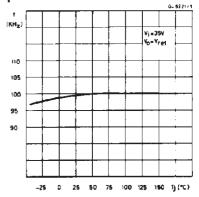


Figure 13. Switching frequency vs. R2 (see test circuit)

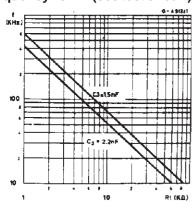


Figure 14. Line transient response

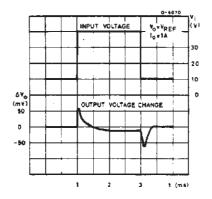
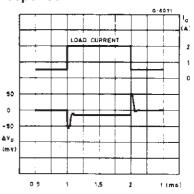


Figure 15. Load transient response



 $\overline{\mathbf{A}}$

Figure 16. Supply voltage ripple rejection vs. frequency

Figure 17. Dropout voltage between pin 1 and pin 7 vs. current at pin 7

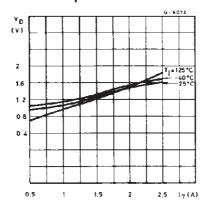


Figure 18. Dropout voltage between pin 1 and 7 vs. junction temperature

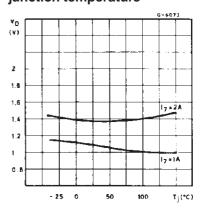


Figure 19. Power dissipation derating curve

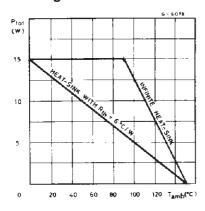


Figure 20. Efficiency vs. output current

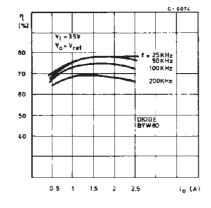


Figure 21. Efficiency vs. output current

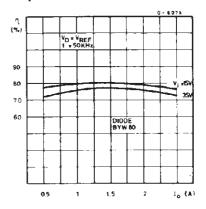


Figure 22. Efficiency vs. output current

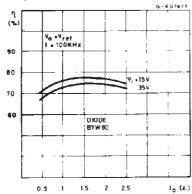
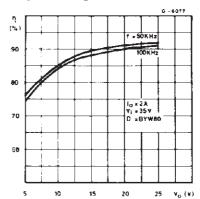


Figure 23. Efficiency vs. output voltage



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APPLICATION INFORMATION

Figure 24. Typical application circuit

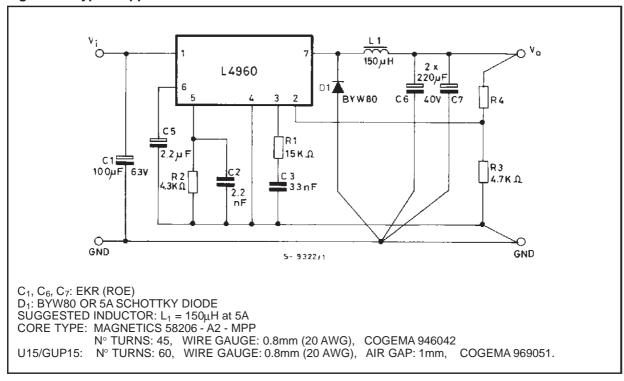
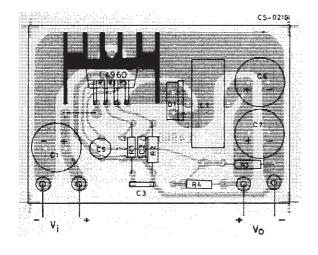


Figure 25. P.C. board and component layout of the Fig. 24 (1: 1 scale)



Resistor values for standard output voltages									
V _o R3 R4									
12V	4.7ΚΩ	6.2KΩ							
15V 18V	4.7KΩ 4.7KΩ	9.1KΩ 12KΩ							
24V									

APPLICATION INFORMATION

Figure 26. A minimal 5.1V fixed regulator; Very few component are required

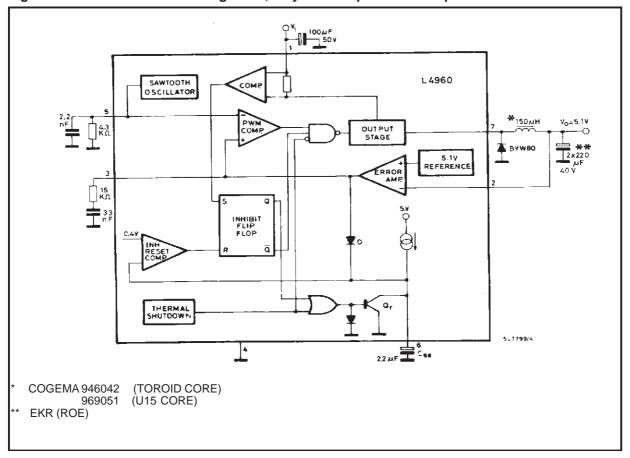
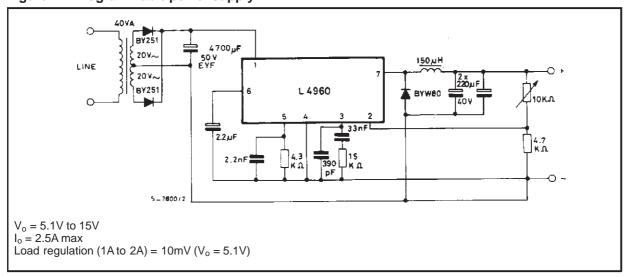
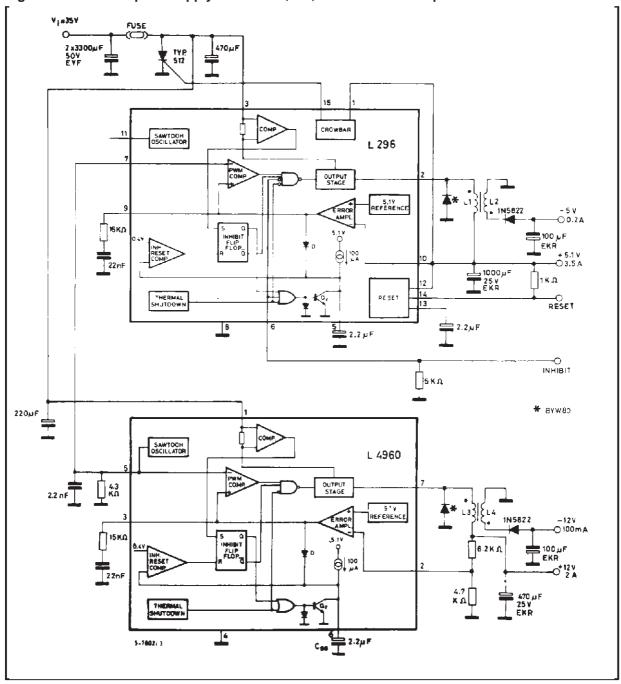


Figure 27. Programmable power supply



APPLICATION INFORMATION (continued)

Figure 28. Microcomputer supply with + 5.1V, -5V, +12V and -12V outputs



APPLICATION INFORMATION (continued)

Figure 29. DC-DC converter $5.1V/4A, \pm 12V/2.5A$; a suggestion how to synchronize a negative output

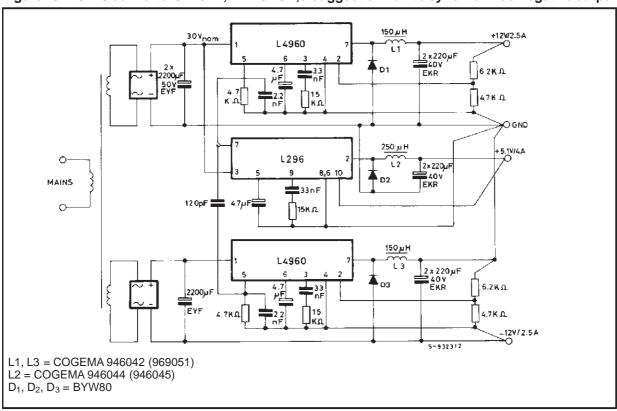
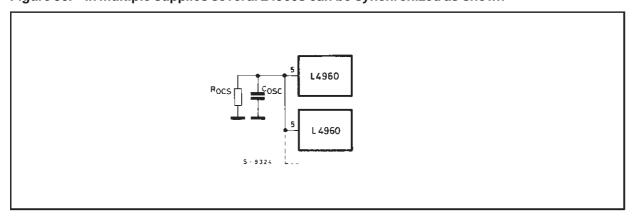
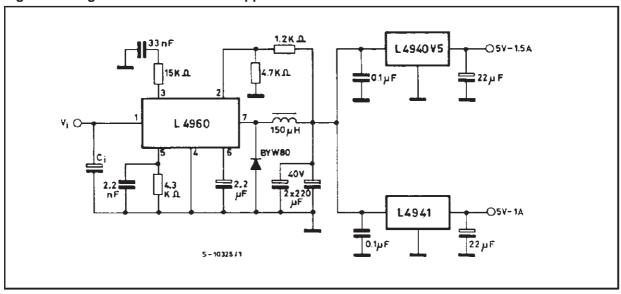


Figure 30. - In multiple supplies several L4960s can be synchronized as shown



APPLICATION INFORMATION (continued)

Figure 31. Regulator for distributed supplies



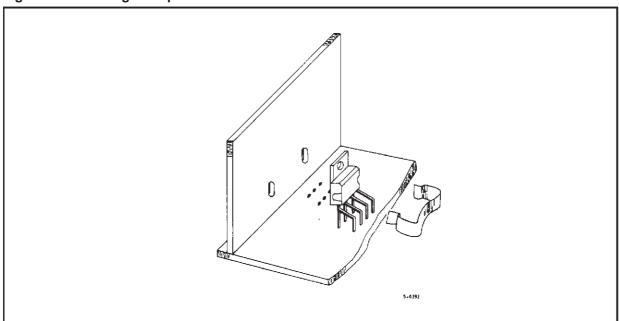
MOUNTING INSTRUCTION

The power dissipated in the circuit must be removed by adding an external heatsink.

Thanks to the Heptawatt package attaching the hetsink is very simple, a screw or a compression spring (clip) being sufficient. Between the heatsink

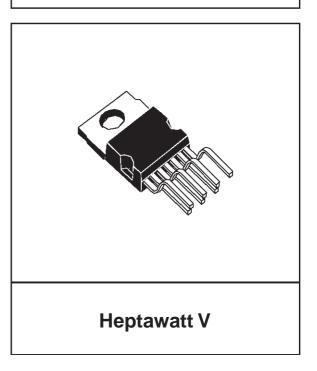
and the package it is better to insert a layer of silicon grease, to optimize the thermal contact, no electrical isolation is needed between the two surfaces.

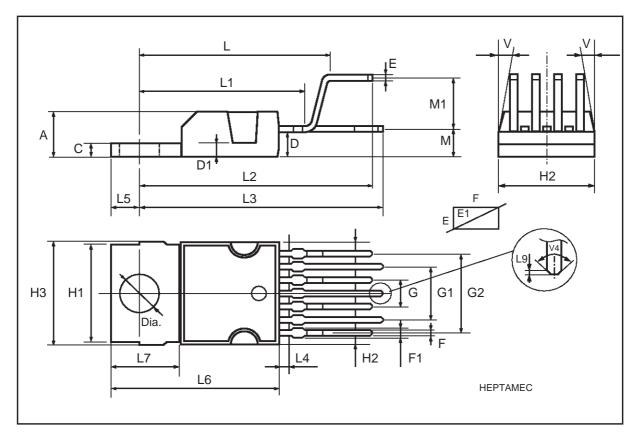
Figure 32. Mounting example



		mm			inch	
DIM.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
Α			4.8			0.189
С			1.37			0.054
D	2.4		2.8	0.094		0.110
D1	1.2		1.35	0.047		0.053
E	0.35		0.55	0.014		0.022
E1	0.7		0.97	0.028		0.038
F	0.6		0.8	0.024		0.031
F1			0.9			0.035
G	2.34	2.54	2.74	0.095	0.100	0.105
G1	4.88	5.08	5.28	0.193	0.200	0.205
G2	7.42	7.62	7.82	0.295	0.300	0.307
H2			10.4			0.409
H3	10.05		10.4	0.396		0.409
L	16.7	16.9	17.1	0.657	0.668	0.673
L1		14.92			0.587	
L2	21.24	21.54	21.84	0.386	0.848	0.860
L3	22.27	22.52	22.77	0.877	0.891	0.896
L4			1.29			0.051
L5	2.6	2.8	3	0.102	0.110	0.118
L6	15.1	15.5	15.8	0.594	0.610	0.622
L7	6	6.35	6.6	0.236	0.250	0.260
L9		0.2			0.008	
М	2.55	2.8	3.05	0.100	0.110	0.120
M1	4.83	5.08	5.33	0.190	0.200	0.210
V4			40°	(typ.)		
Dia	3.65		3.85	0.144		0.152

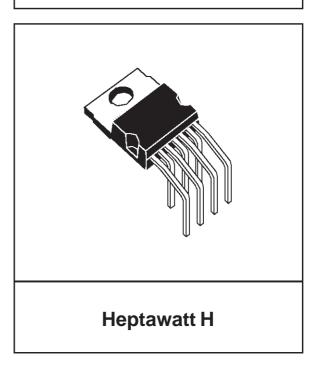
OUTLINE AND MECHANICAL DATA

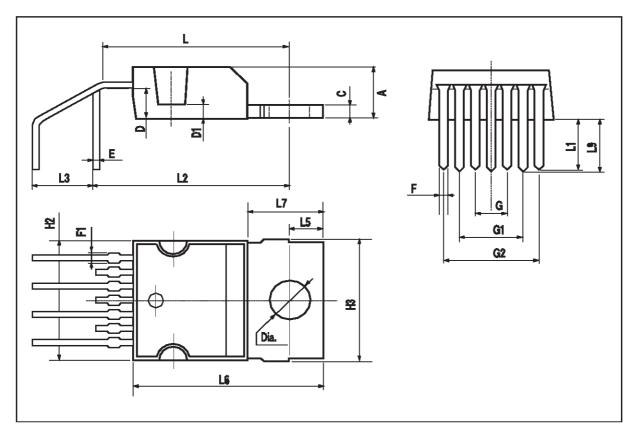




DIM.	mm			inch		
Dilvi.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
Α			4.8			0.189
С			1.37			0.054
D	2.4		2.8	0.094		0.110
D1	1.2		1.35	0.047		0.053
E	0.35		0.55	0.014		0.022
F	0.6		0.8	0.024		0.031
F1			0.9			0.035
G	2.41	2.54	2.67	0.095	0.100	0.105
G1	4.91	5.08	5.21	0.193	0.200	0.205
G2	7.49	7.62	7.8	0.295	0.300	0.307
H2			10.4			0.409
НЗ	10.05		10.4	0.396		0.409
L		14.2			0.559	
L1		4.4			0.173	
L2		15.8			0.622	
L3		5.1			0.201	
L5	2.6		3	0.102		0.118
L6	15.1		15.8	0.594		0.622
L7	6		6.6	0.236		0.260
L9		4.44			0.175	
Dia	3.65		3.85	0.144		0.152

OUTLINE AND MECHANICAL DATA





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